

### General Description

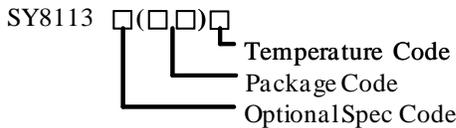
The SY8113G is a high efficiency 500 kHz synchronous step-down DC-DC converter capable of delivering 3A current. The SY8113G operates over a wide input voltage range from 4.5V to 18V and integrates main switch and synchronous switch with very low  $R_{DS(ON)}$  to minimize the conduction loss.

SY8113G always operates under continuous condition mode, low output ripple and small inductor and capacitor size are achieved with 500kHz switching frequency.

### Features

- low  $R_{DS(ON)}$  for internal switches (top/bottom): 80mΩ/40mΩ
- 4.5-18V input voltage range
- 3A output current capability
- 500 kHz switching frequency
- Instant PWM architecture to achieve fast transient responses.
- Cycle-by-cycle peak current limitation
- Internal softstart limits the inrush current
- $\pm 1.5\%$  0.6V reference
- TSOT23-6 package

### Ordering Information



Ordering Number	Package type	Note
SY8113GADC	TSOT23-6	--

### Applications

- Set Top Box
- Portable TV
- Access Point Router
- DSL Modem
- LCD TV

### Typical Applications

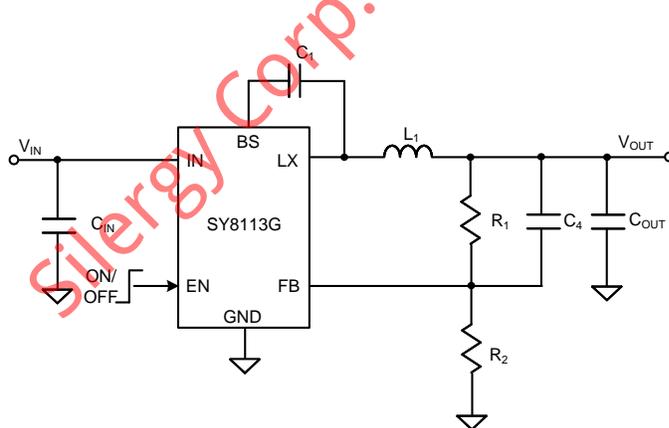


Figure 1. Schematic Diagram

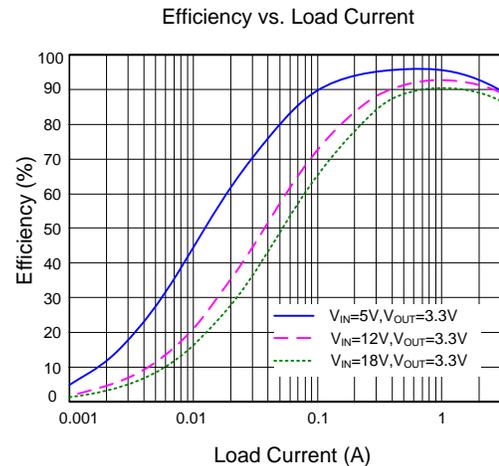
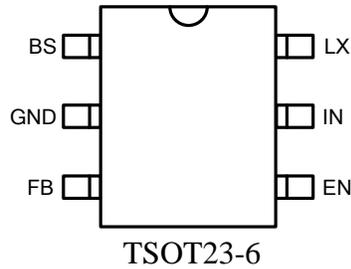


Figure 2. Efficiency vs. Load Current

## Pinout (top view)



Top Mark: Itxyz, (Device code: It, *x*=year code, *y*=week code, *z*=lot number code)

Pin Name	Pin Number	Pin Description
BS	1	Boot-Strap Pin. Supply high side gate driver. Decouple this pin to LX pin with 0.1uF ceramic cap.
GND	2	Ground pin
FB	3	Output Feedback Pin. Connect this pin to the center point of the output resistor divider (as shown in Figure 1) to program the output voltage: $V_{out}=0.6*(1+R1/R2)$
EN	4	Enable control. Pull high to turn on. Do not float.
IN	5	Input pin. Decouple this pin to GND pin with at least 1uF ceramic cap
LX	6	Inductor pin. Connect this pin to the switching node of inductor

## Absolute Maximum Ratings (Note 1)

Supply Input Voltage	19V
Enable Voltage	$V_{IN} + 0.3V$
FB Voltage	4V
Power Dissipation, PD @ $T_A = 25^\circ C$ , TSOT23-6	1.5W
Package Thermal Resistance (Note 2)	
$\theta_{JA}$	66 $^\circ C/W$
$\theta_{JC}$	15 $^\circ C/W$
Junction Temperature Range	150 $^\circ C$
Lead Temperature (Soldering, 10 sec.)	260 $^\circ C$
Storage Temperature Range	-65 $^\circ C$ to 150 $^\circ C$

## Recommended Operating Conditions (Note 3)

Supply Input Voltage	4.5V to 18V
Junction Temperature Range	-40 $^\circ C$ to 125 $^\circ C$
Ambient Temperature Range	-40 $^\circ C$ to 85 $^\circ C$



**Electrical Characteristics**

( $V_{IN} = 12V$ ,  $V_{OUT} = 1.2V$ ,  $L = 2.2\mu H$ ,  $C_{OUT} = 47\mu F$ ,  $T_A = 25\text{ }^\circ C$ ,  $I_{OUT} = 1A$  unless otherwise specified)

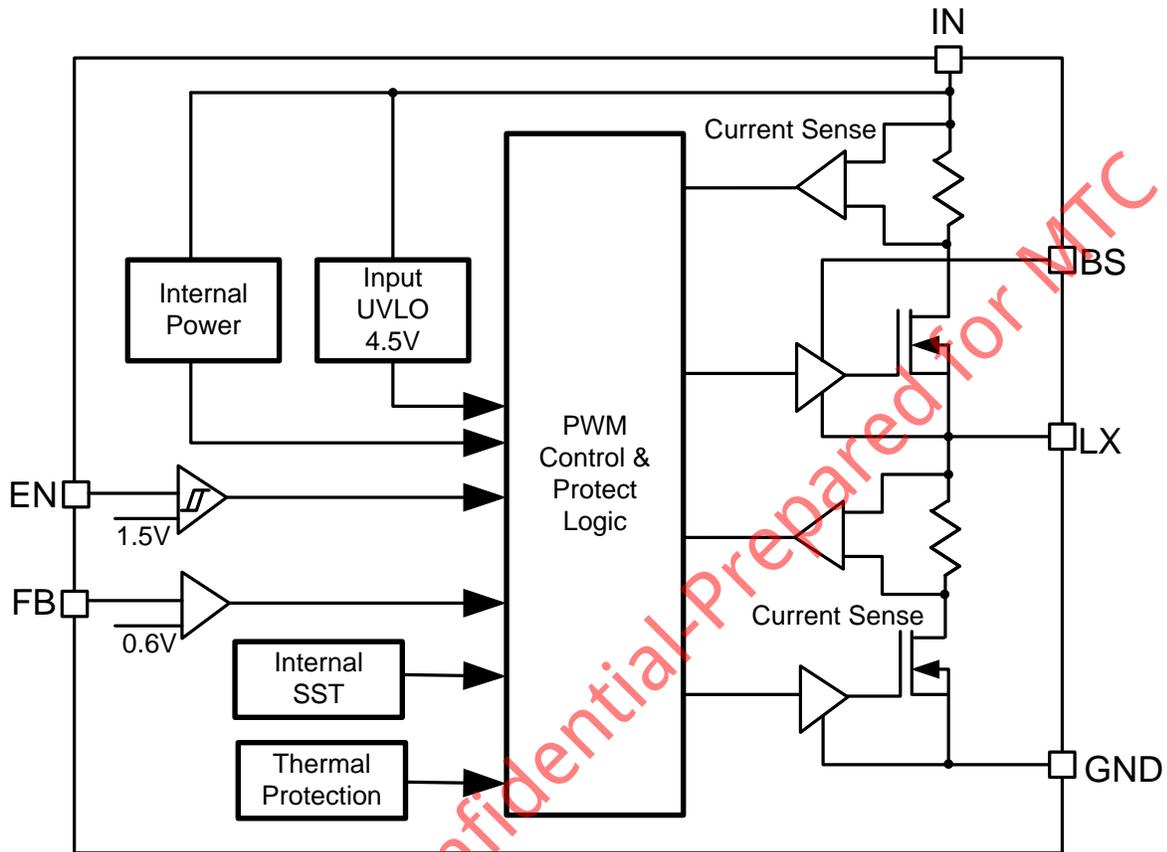
Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Input Voltage Range	$V_{IN}$		4.5		18	V
Shutdown Current	$I_{SHDN}$	EN=0		5	10	$\mu A$
Feedback Reference Voltage	$V_{REF}$		0.591	0.6	0.609	V
FB Input Current	$I_{FB}$	$V_{FB}=3.3V$	-50		50	nA
Top FET RON	$R_{DS(ON)1}$			80	90	m $\Omega$
Bottom FET RON	$R_{DS(ON)2}$			40	50	m $\Omega$
Top FET Peak Current Limit	$I_{LIM, TOP}$		5.8		6.6	A
Bottom FET Valley Current Limit	$I_{LIM}$		4		5.2	A
EN Rising Threshold	$V_{ENH}$		1.5			V
EN Falling Threshold	$V_{ENL}$				0.4	V
Input UVLO Threshold	$V_{UVLO}$		3.8	4.2	4.5	V
UVLO Hysteresis	$V_{HYS}$			0.3		V
Min ON Time			40	60	100	ns
Min OFF Time			140	170	220	ns
Switching Frequency				500		kHz
Soft-start Time	$t_{SS}$			800		$\mu S$
Thermal Shutdown Temperature	$T_{SD}$			150		$^\circ C$
Thermal Shutdown Hysteresis	$T_{HYS}$			15		$^\circ C$

**Note 1:** Stresses beyond the “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**Note 2:**  $\theta_{JA}$  is measured in the natural convection at  $T_A = 25\text{ }^\circ C$  on a two-layer Silergy Evaluation Board.

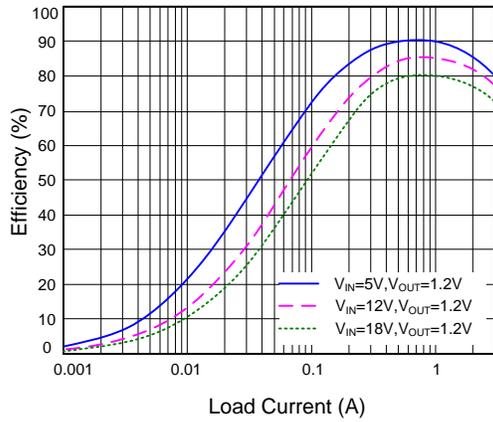
**Note 3:** The device is not guaranteed to function outside its operating conditions

**Block Diagram**

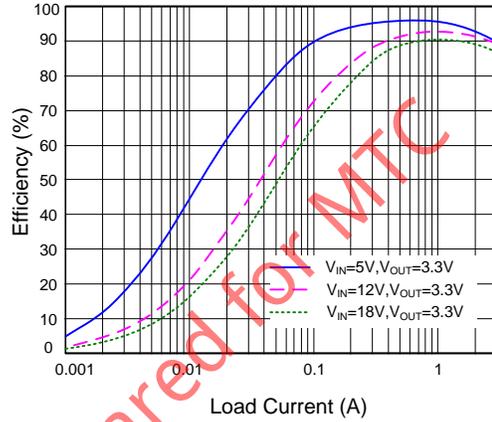


## Typical Performance Characteristics

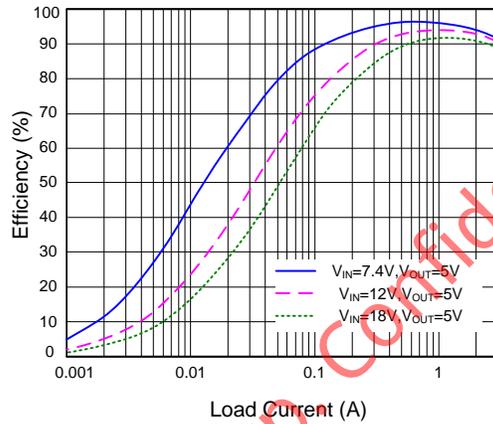
Efficiency vs. Load Current



Efficiency vs. Load Current

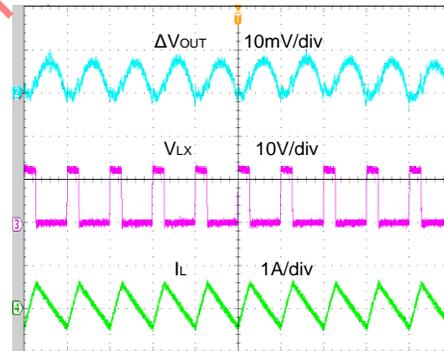


Efficiency vs. Load Current



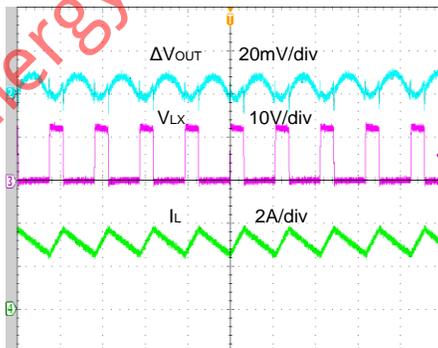
Output Ripple

( $V_{IN}=12V, V_{OUT}=3.3V, I_{OUT}=0A$ )



Output Ripple

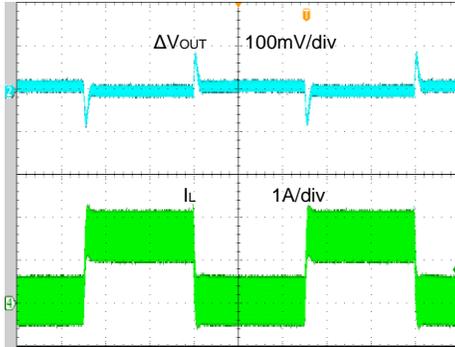
( $V_{IN}=12V, V_{OUT}=3.3V, I_{OUT}=3A$ )



Time (2 $\mu$ s/div)

### Load Transient

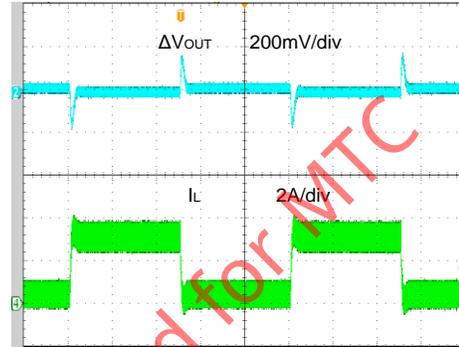
( $V_{IN}=12V$ ,  $V_{OUT}=3.3V$ ,  $I_{OUT}=0-1.5A$ )



Time (200µs/div)

### Load Transient

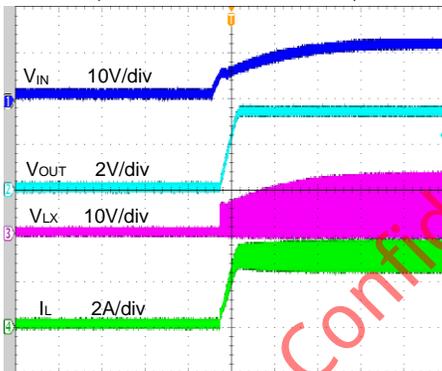
( $V_{IN}=12V$ ,  $V_{OUT}=3.3V$ ,  $I_{OUT}=0.3-3A$ )



Time (200µs/div)

### Startup from VIN

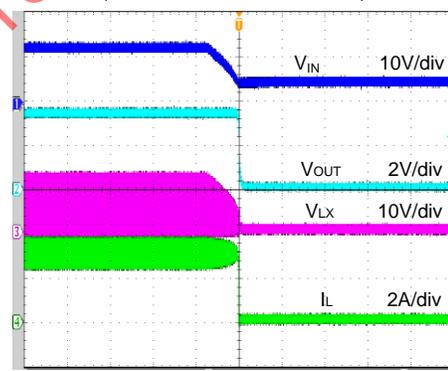
( $V_{IN}=12V$ ,  $V_{OUT}=3.3V$ ,  $I_{OUT}=3A$ )



Time (2ms/div)

### Shutdown from VIN

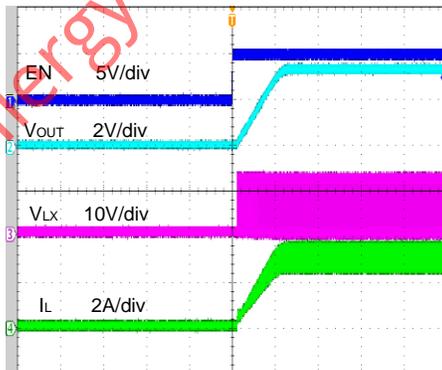
( $V_{IN}=12V$ ,  $V_{OUT}=3.3V$ ,  $I_{OUT}=3A$ )



Time (2ms/div)

### Startup from Enable

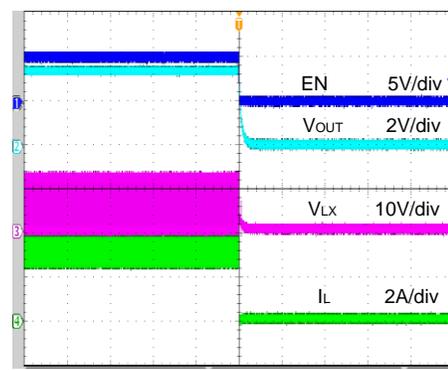
( $V_{IN}=12V$ ,  $V_{OUT}=3.3V$ ,  $I_{OUT}=3A$ )



Time (800µs/div)

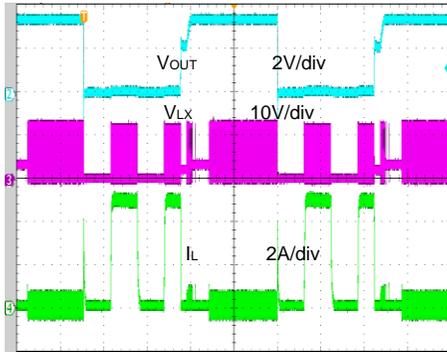
### Shutdown from Enable

( $V_{IN}=12V$ ,  $V_{OUT}=3.3V$ ,  $I_{OUT}=3A$ )



Time (800µs/div)

Short Circuit Protection  
 ( $V_{IN}=12V$ ,  $V_{OUT}=3.3V$ , 0A to Short)



Time (4ms/div)

Short Circuit Protection  
 ( $V_{IN}=12V$ ,  $V_{OUT}=3.3V$ , 3A to Short)



Time (4ms/div)

## Operation

The SY8113G is a high efficiency 500 kHz synchronous step-down DC-DC converter capable of delivering 3A current. The SY8113G operates over a wide input voltage range from 4.5V to 18V and integrates main switch and synchronous switch with very low  $R_{DS(ON)}$  to minimize the conduction loss.

SY8113G always operates under continuous condition mode, low output ripple and small inductor and capacitor size are achieved with 500kHz switching frequency.

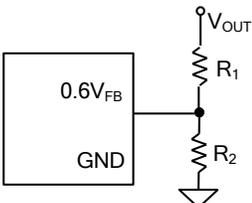
SY8113G provides protection functions such as cycle by cycle current limiting and thermal shutdown protection. SY8113G will sense the output voltage conditions for the fault protection.

## Applications Information

Because of the high integration in the SY8113G IC, the application circuit based on this regulator IC is rather simple. Only input capacitor  $C_{IN}$ , output capacitor  $C_{OUT}$ , output inductor  $L$  and feedback resistors ( $R_1$  and  $R_2$ ) need to be selected for the targeted applications specifications.

### Feedback resistor dividers $R_1$ and $R_2$ :

Choose  $R_1$  and  $R_2$  to program the proper output voltage. To minimize the power consumption under light loads, it is desirable to choose large resistance values for both  $R_1$  and  $R_2$ . A value of between 10k $\Omega$  and 1M $\Omega$  is highly recommended for both resistors. If  $V_{out}$  is 3.3V,  $R_1=100k$  is chosen, then using following equation,  $R_2$  can be calculated to be 22.1k:

$$R_2 = \frac{0.6V}{V_{OUT} - 0.6V} R_1$$


### Input capacitor $C_{IN}$ :

The ripple current through input capacitor is calculated as:

$$I_{CIN\_RMS} = I_{OUT} \cdot \sqrt{D(1-D)}$$

To minimize the potential noise problem, place a typical X5R or better grade ceramic capacitor really close to the IN and GND pins. Care should be taken to minimize the loop area formed by  $C_{IN}$ , and IN/GND pins. In this case, a 10uF low ESR ceramic capacitor is recommended.

### Output capacitor $C_{OUT}$ :

The output capacitor is selected to handle the output ripple noise requirements. Both steady state ripple and transient requirements must be taken into consideration when selecting this capacitor. For the best performance, it is recommended to use X5R or better grade ceramic capacitor greater than 22uF capacitance.

### Output inductor $L$ :

There are several considerations in choosing this inductor.

- 1) Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the maximum output current. The inductance is calculated as:

$$L = \frac{V_{OUT}(1 - V_{OUT}/V_{IN,MAX})}{F_{SW} \times I_{OUT,MAX} \times 40\%}$$

where  $F_{sw}$  is the switching frequency and  $I_{OUT,MAX}$  is the maximum load current.

The SY8113G regulator IC is quite tolerant of different ripple current amplitude. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

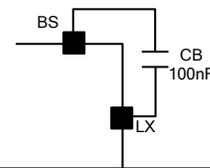
- 2) The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load conditions.

$$I_{SAT, MIN} > I_{OUT, MAX} + \frac{V_{OUT}(1 - V_{OUT}/V_{IN,MAX})}{2 \cdot F_{SW} \cdot L}$$

- 3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with  $DCR < 50m\Omega$  to achieve a good overall efficiency.

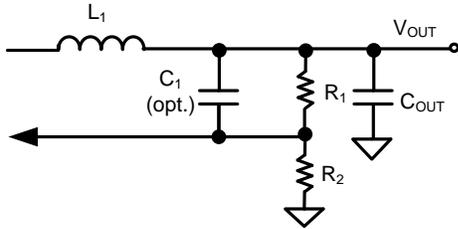
### External Bootstrap Cap

This capacitor provides the gate driver voltage for internal high side MOSEFET. A 100nF low ESR ceramic capacitor connected between BS pin and LX pin is recommended.



**Load Transient Considerations:**

The SY8113G regulator IC integrates the compensation components to achieve good stability and fast transient responses. In some applications, adding a 22pF ceramic cap in parallel with R1 may further speed up the load transient responses and is thus recommended for applications with large load transient step requirements.



**Layout Design:**

The layout design of SY8113G regulator is relatively simple. For the best efficiency and minimum noise problem, we should place the following components close to the IC: CIN, L, R1 and R2.

1) It is desirable to maximize the PCB copper area connecting to GND pin to achieve the best thermal and

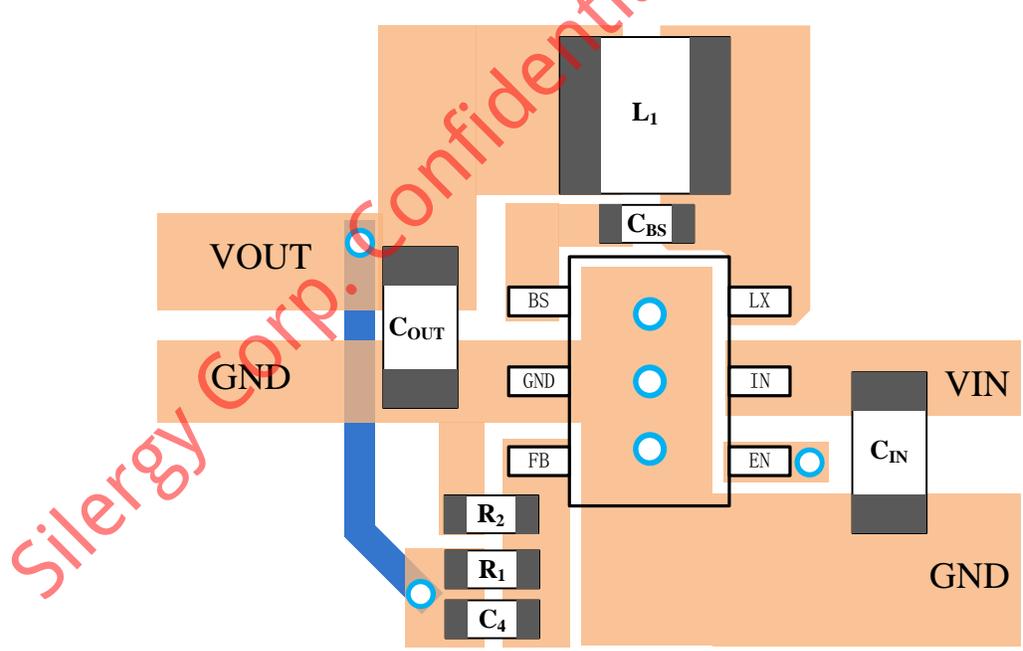
noise performance. If the board space allowed, a ground plane is highly desirable.

2) CIN must be close to Pins IN and GND. The loop area formed by CIN and GND must be minimized.

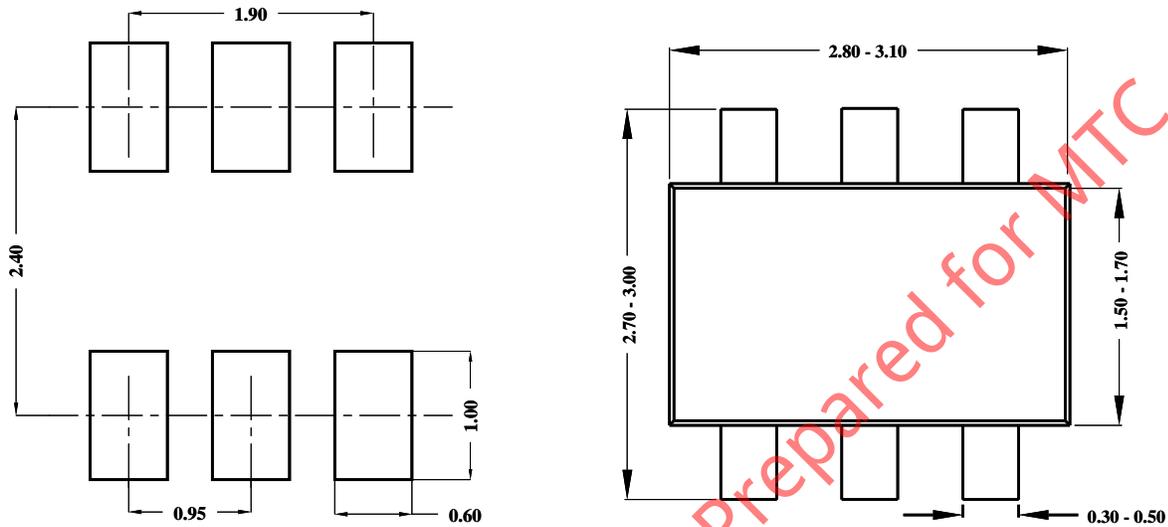
3) The PCB copper area associated with LX pin must be minimized to avoid the potential noise problem.

4) The components R1 and R2, and the trace connecting to the FB pin must NOT be adjacent to the LX net on the PCB layout to avoid the noise problem.

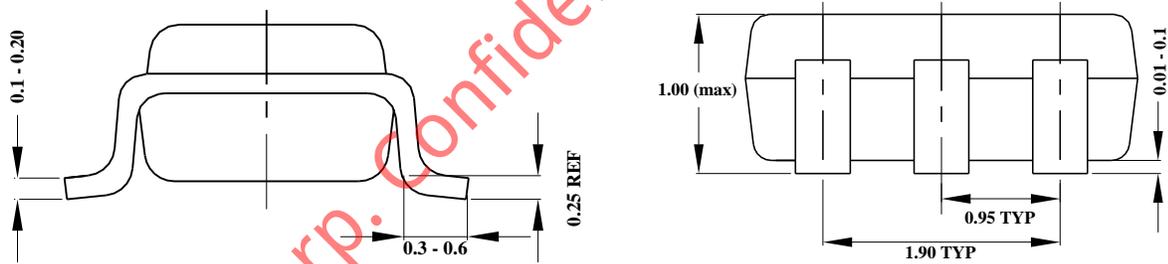
5) If the system chip interfacing with the EN pin has a high impedance state at shutdown mode and the IN pin is connected directly to a power source such as a Li-Ion battery, it is desirable to add a pull down 1Mohm resistor between the EN and GND pins to prevent the noise from falsely turning on the regulator at shutdown mode.



## TSOT23-6 Package outline & PCB layout



### Recommended Pad Layout

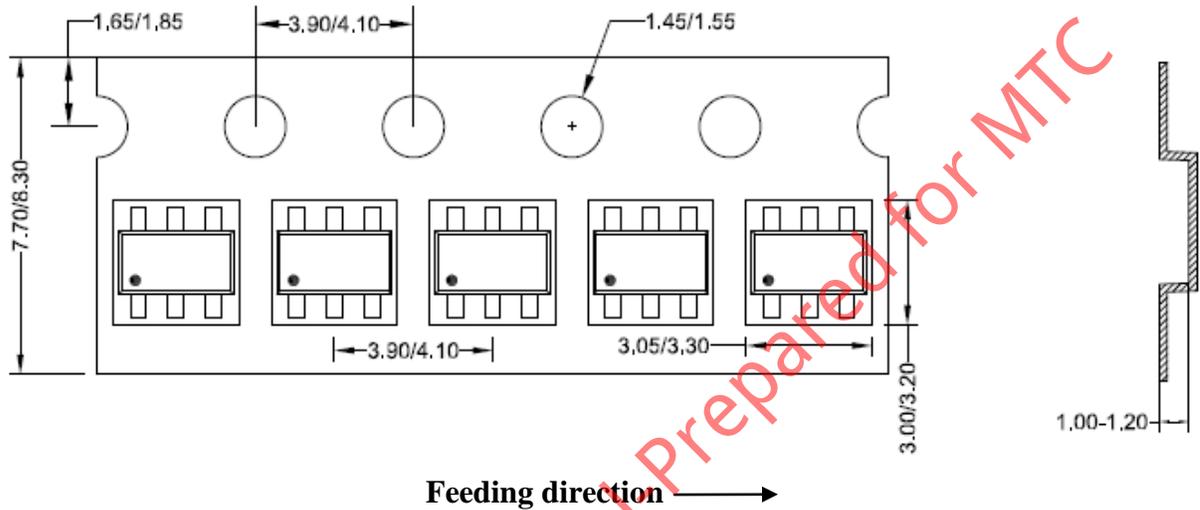


**Notes:** All dimension in MM  
 All dimension don't include mold flash & metal burr

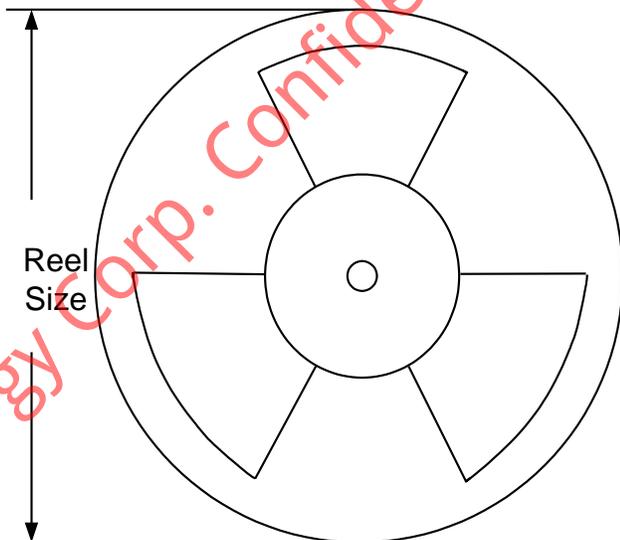
**Taping & Reel Specification**

**1. Taping orientation**

TSOT23-6



**2. Carrier Tape & Reel specification for packages**



Package types	Tape width (mm)	Pocket pitch(mm)	Reel size (Inch)	Trailer length(mm)	Leader length (mm)	Qty per reel
TSOT23-6	8	4	7	400	160	3000

**3. Others: NA**



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