



# TMS8008 - Low EMI 3.0W Mono Audio Amplifier

## **FEATURES**

- Supply Voltage from 2.5V to 6.0V
- 3.0W@10% THD Output with a 4Ω Load at 5V
  Supply
- High Efficiency Up to 90% @1W with an 8Ω
  Speaker
- Shutdown Current <1µA</p>
- Superior Low Noise without Input
- EMI Suppressing by Soft-Driving
- Short Circuit Protection
- Thermal Shutdown
- Available in Space Saving DFN2x2-8L and MSOP-8L Package

### GENERAL DESCRIPTION

The TMS8008 is a mono filter-less class-D amplifier with high SNR and differential input that eliminate noise.

Features like higher than 90% efficiency and small PCB areas make the TMS8008 class-D amplifier ideal for portable devices. The filter-less architecture requires no external output filter, fewer external components, less PCB areas and lower system costs, and simplifies application design.

With the soft-driving technology, the edge of the PWM at output stage is very flat which is very useful for EMI suppressing.

The TMS8008 features short circuit protection, thermal shutdown and under voltage lock-out. The TMS8008 is available in DFN2x2-8L and MSOP-8L packages.

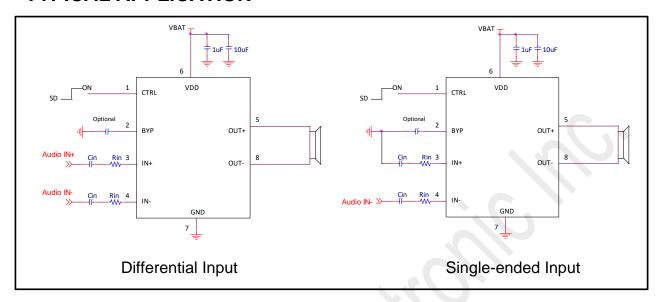
## **APPLICATIONS**

- Wearable Device
- Cellphones
- Telephone Watches
- IPCs
- Portable Device
- AIOT





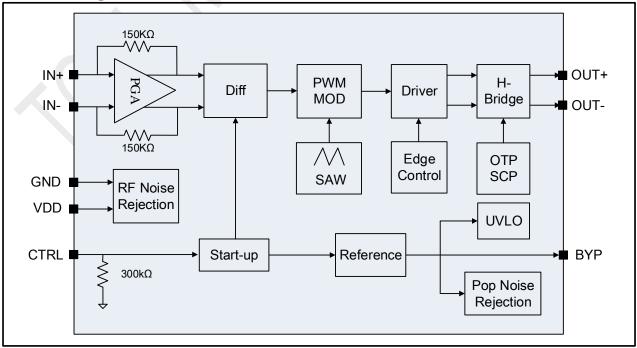
# **TYPICAL APPLICATION**



### **ABSOLUTE MAXIMUM RATINGS**

Parameter	Rating	Unit
V <sub>DD</sub> Supply Voltage	6.5	V
Minimum Output Impedance	3.0	Ω
Input Voltage (IN+, IN-, CTRL)	-0.3 to V <sub>DD</sub> +0.3	V
Storage Temperature	-65 to 150	°C
Maximum Junction Temperature	150	°C

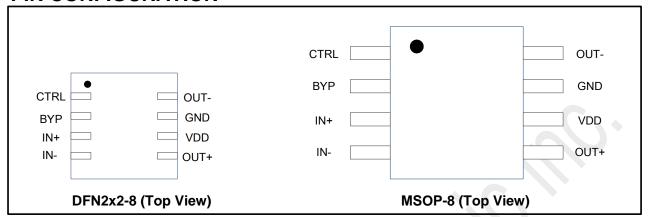
# **Block Diagram**







# PIN CONFIGURATION



### **PIN FUNCTIONS**

Pin Number	Pin name	Description		
1	CTRL	Control Terminal, Pull Low Internally		
2	ВҮР	Internal Reference Voltage Bypass Pin; Connect a 1.0uF		
2	DIF	Capacitance from Thins Pin to GND		
3	IN+	Positive Differential Input		
4	IN-	Negative Differential Input		
5	OUT-	Negative BTL Output		
6	V <sub>DD</sub>	Power Supply		
7	GND	Ground		
8	OUT+	Positive BTL Output		

# **PART NUMBER & MARKING**

Part Number	Package	Top Mark	Description	Quantity/ Reel	
TMC0000LID TD		TA2x	TA2: Device Code	2000	
TMS8008UP-TR	DFN2x2-8	xxxx	xxxx: Internal Code	3000	
TMC000CD TD MCOD 0		T8008SP	T8008SP: Device Code	2000	
TMS8008SP-TR	MSOP-8	xxxxxx	xxxxx: Internal Code	3000	

TMS8020 devices are Pb-free and RoHS compliant.

## **ESD RATING**

Items	Description	Value	Unit
V <sub>ESD_HBM</sub>	Human Body Model	±4000	V
V <sub>ESD_CDM</sub>	Charge Device Model	±750	V

**JEDEC specification JS-001** 







# **RECOMMENDED OPERATING CONDITIONS**

Symbol	Parameter	Min	Max	Unit
$V_{DD}$	Supply Voltage	2.5	6.0	V
Ta	Operating Ambient Temperature Range	-25	85	°C
TJ	Junction Temperature Range	-40	125	°C

## **ELECTRICAL CHARACTERISTICS**

 $(T_a=25^{\circ}C, V_{DD}=5V, R_{IN}=20K\Omega, C_{IN}=0.22uF, R_L=L(33\mu H) + R+L(33\mu H), unless otherwise noted.)$ 

Symbol	Parameter	Test Conditi	ons	MIN	TYP	MAX	UNIT	
		THD+N=10%, f=1kHz,	V <sub>DD</sub> =5.0V		3.0	3.0	14/	
	R <sub>L</sub> =4Ω	V <sub>DD</sub> =3.7V		1.65		W		
		THD+N=1%, f=1kHz,	V <sub>DD</sub> =5.0V		2.55		W	
D-	Outrast Bassas	$R_L = 4\Omega$	V <sub>DD</sub> =3.7V		1.36			
Po	Output Power	THD+N=10%, f=1kHz,	V <sub>DD</sub> =5.0V		1.72		14/	
		R <sub>L</sub> =8Ω	V <sub>DD</sub> =3.7V		0.94		W	
		THD+N=1%, f=1kHz,	V <sub>DD</sub> =5.0V		1.40		147	
		R <sub>L</sub> =8Ω	V <sub>DD</sub> =3.7V		0.75		W	
		V <sub>DD</sub> =5.0V, Po=0.25W	f=1kHz,		0.036		%	
TUD. N	Total Harmonic	V <sub>DD</sub> =3.7V, Po=0.25W	R <sub>L</sub> =8Ω		0.015			
THD+N	Distortion Plus Noise	V <sub>DD</sub> =5.0V, Po=0.5W	f=1kHz,		0.02		%	
		V <sub>DD</sub> =3.7V, Po=0.5W	R <sub>L</sub> =4Ω		0.026			
PSRR	Power Supply Ripple	V <sub>DD</sub> =5V, Inputs AC-	f=217Hz		-75		dB	
PSKK	Rejection	Grounded, CBYP=0.1uF	f=1kHz		-75			
SNR	Signal-to-Noise Ratio	THD=1%, f=1kHz	A-weighting		96		dB	
Vn	Output Noise	Inputs AC-Grounded,	No A- weighting		55		μV	
	o a par i roico	Gv=6dB	A-weighting		51		μv	
G∨	Closed-loop Gain	V <sub>DD</sub> = 5V			300K/Rin		V/V	
fsw	Switching Frequency	V <sub>DD</sub> = 5V			590		kHz	
	T#inings.	R <sub>L</sub> =8Ω, THD=10%	4 4141-		91		0/	
η	Efficiency	R <sub>L</sub> =4Ω, THD=10%	f=1kHz		87		%	
	0.1	V <sub>DD</sub> =5V	Natard		2.2			
lq	Quiescent Current	V <sub>DD</sub> =3.7V	No Load		1.7		mA	
DC Parar	neters							
I <sub>SD</sub>	Shutdown Current	V <sub>DD</sub> =5V	CTRL=0V			1	μA	
RSDON	Static Drain-to Source On-state Resistor	High Side + Low Side	V <sub>DD</sub> =5.0V, I=500mA		400		mΩ	
Ton	Turn On Time	V <sub>DD</sub> = 5V	C <sub>BYP</sub> =0.1uF		32		mS	
Vos	Output Offset Voltage	V <sub>DD</sub> =5V	AC-Ground		3.5		mV	
VIH	Input High Voltage	V <sub>DD</sub> =5V		1.4			V	
VIL	Input Low Voltage	V <sub>DD</sub> =5V				1.0	V	

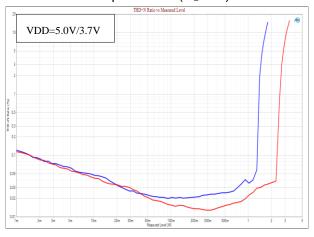




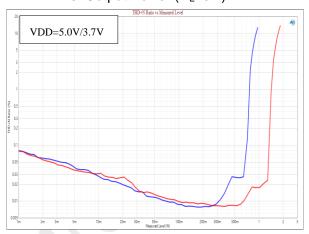
## PERFORMANCE CHARACTERISTICS

 $(T_a=25^{\circ}C,\ V_{DD}=5V,\ R_{IN}=20K\Omega,\ C_{IN}=0.22uF,\ R_L=L(33\mu H)+R+L(33\mu H),\ Differential\ Input,\ unless\ otherwise\ noted.)$ 

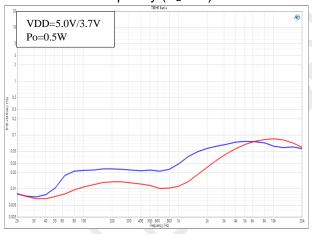
### THD+N Vs. Output Power ( $R_L$ = $4\Omega$ )



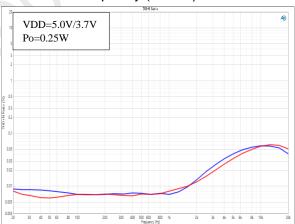
### THD+N Vs. Output Power ( $R_L=8\Omega$ )



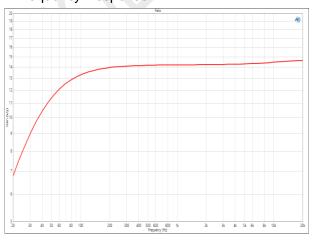
THD+N Vs. Frequency ( $R_L=4\Omega$ )



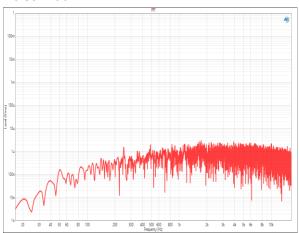
THD+N Vs. Frequency ( $R_L=8\Omega$ )



### Frequency Response



#### Noise Floor



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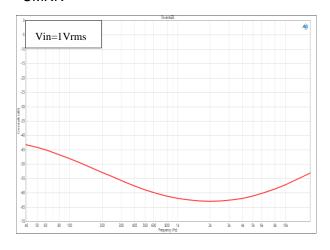




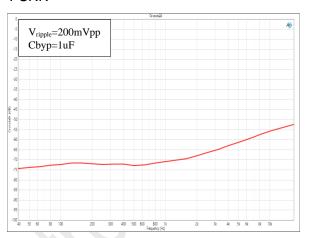
## PERFORMANCE CHARACTERISTICS

 $(T_a=25^{\circ}C,\,V_{DD}=5V,\,R_{IN}=20K\Omega,\,C_{IN}=0.22uF,\,R_L=L(33\mu H)\,+\,R+L(33\mu H),\,unless\,\,otherwise\,\,noted.)$ 

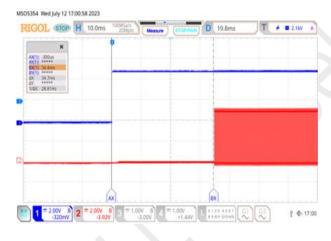
#### **CMRR**



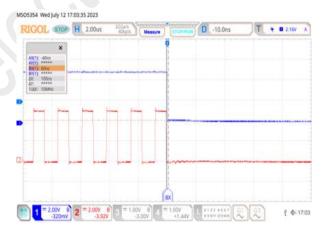
#### **PSRR**



#### Start-up Response



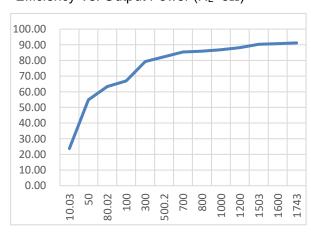
Shutdown Response



### Efficiency Vs. Output Power ( $R_L=4\Omega$ )



Efficiency Vs. Output Power (R<sub>L</sub>=8Ω)



Output Power (mW)





### APPLICATION INFORMATION

#### Input Capacitors (Ci)

In the typical application, an input capacitor, Ci, is required to allow the amplifier to bias the input signal to the proper DC level for optimum operation. In this case, Ci and the minimum input impedance Ri form is a high-pass filter with the corner frequency determined in the follow equation:

$$f_{C} = \frac{1}{(2\pi RiCi)}$$

It is important to consider the value of Ci as it directly affects the low frequency performance of the circuit. For example, when Ri is  $160k\Omega$  and the specification calls for a flat bass response are down to 150Hz. Equation is reconfigured as followed:

$$Ci = \frac{1}{(2\pi R_i f_c)}$$

When input resistance variation is considered, the Ci is 7nF, so one would likely choose a value of 10nF. A further consideration for this capacitor is the leakage path from the input source through the input network (Ci, Ri + Rf) to the load. This leakage current creates a DC offset voltage at the input to the amplifier that reduces useful headroom, especially in high gain applications. For this reason, a low-leakage tantalum or ceramic capacitor is the best choice. When polarized capacitors are used, the positive side of the capacitor should face the amplifier input in most applications as the DC level is held at VDD/2, which is likely higher than the source DC level. Please note that it is important to confirm the capacitor polarity in the application.

### **Decoupling Capacitor (CS)**

The TMS8008 is a high-performance CMOS audio amplifier that requires adequate power supply decoupling to ensure the output total harmonic distortion (THD) as low as possible. Power supply decoupling also prevents the oscillations causing by long lead length between the amplifier and the speaker.

The optimum decoupling is achieved by using two different types of capacitors that target on different types of noise on the power supply leads. For higher frequency transients, spikes, or digital hash on the line, a good low Equivalent-Series-Resistance (ESR) ceramic capacitor, typically  $1\mu F$ , is placed as close as possible to the device VDD pin for the best operation. For filtering lower frequency noise signals, a large ceramic capacitor of  $10\mu F$  or greater placed near the audio power amplifier is recommended. Long conducting wires on VDD or & and GND (>1m) will cause big power pumping which may damage the chip, that an additional  $100\mu F$  or greater capacitance placed near the TMS8008 is needed.

#### How to Reduce EMI

Most applications require a ferrite bead filter for EMI elimination shown at Figure 1. The ferrite filter reduces EMI around 1MHz and higher. When selecting a ferrite bead, choose one with high impedance at high frequencies, but low impedance at low frequencies.

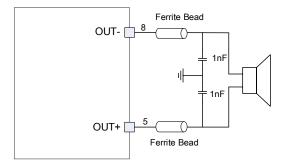


Figure 1: Ferrite Bead Filter to Reduce EMI

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### APPLICATION INFORMATION

### **Under Voltage Lock-out (UVLO)**

The TMS8008 incorporates circuitry designed to detect low supply voltage. When the supply voltage drops to 2.4V or below, the TMS8008 goes into a state of shutdown, and the device comes out of its shutdown state and restore to normal function only when VDD higher than 2.5V.

#### **Short Circuit Protection (SCP)**

The TMS8008 has short circuit protection circuitry on the outputs to prevent the device from damage when output-to-output shorts or output-to-GND shorts occur. When a short circuit occurs, the device immediately goes into shutdown state. Once the short is removed, the device will be reactivated.

#### **Over Temperature Protection (OTP)**

Thermal protection on the TMS8008 prevents the device from damage when the internal die temperature exceeds 135°C. There is a 15°C tolerance on this trip point from device to device. Once the die temperature exceeds the set point, the device will enter the shutdown state and the outputs are disabled. This is not a latched fault. The thermal fault is cleared once the temperature of the die decreased by 15°C. This large hysteresis will prevent motor boating sound well and the device begins normal operation at this point with no external system interaction.

#### **Shutdown Operation**

In order to reduce power consumption while not in use, the TMS8008 contains shutdown circuitry amplifier off when logic low is placed on the CTRL pin. By switching the CTRL pin connected to GND, the TMS8008 supply current draw will be minimized in idle mode.

#### **POP and Click Circuitry**

The TMS8008 contains circuitry to minimize turn-on and turn-off transients or "click and pops", where turn-on refers to either power supply turn-on or device recover from shutdown mode. When the device is turned on, the amplifiers are internally muted. An internal current source ramps up the internal reference voltage. The device will remain in mute mode until the reference voltage reach half supply voltage, 1/2 VDD. As soon as the reference voltage is stable, the device will begin full operation. For the best power-off pop performance, the amplifier should be set in shutdown mode prior to removing the power supply voltage.

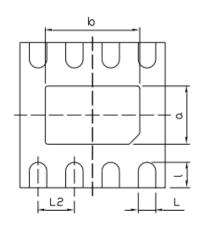
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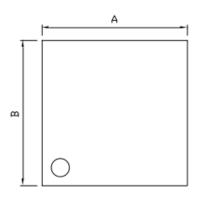




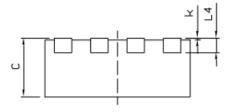
# **PACKAGE INFORMATION**

# DFN2X2-8L

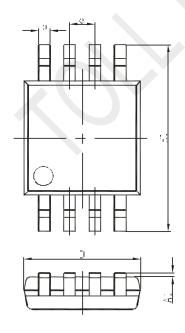


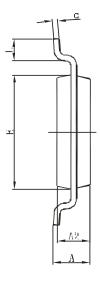


Dimensions In Millimeterer				
Symbol	MIN	TYP	MAX	
Α	1.95	2.00	2.05	
В	1.95	2.00	2.05	
С	0.70	0.75	0.80	
┙	0.19	0.24	0.29	
L2	ı	0.50	-	
L4	ı	0.203	-	
۵	0.75	0.80	0.85	
b	1.25	1.30	1.35	
l	0.30	0.35	0.40	
ĸ	0.00	-	0.05	



## MSOP-8L





REF	Millimeter		
KEF	Min	Max	
Α	-	1.10	
A1	0.05	0.15	
A2	0.78	0.94	
b	0.22	0.38	
С	0.08	0.23	
D	2.90	3.10	
Е	2.90	3.10	
E1	4.75	5.05	
е	0.65BSC		
L	0.40	0.70	





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